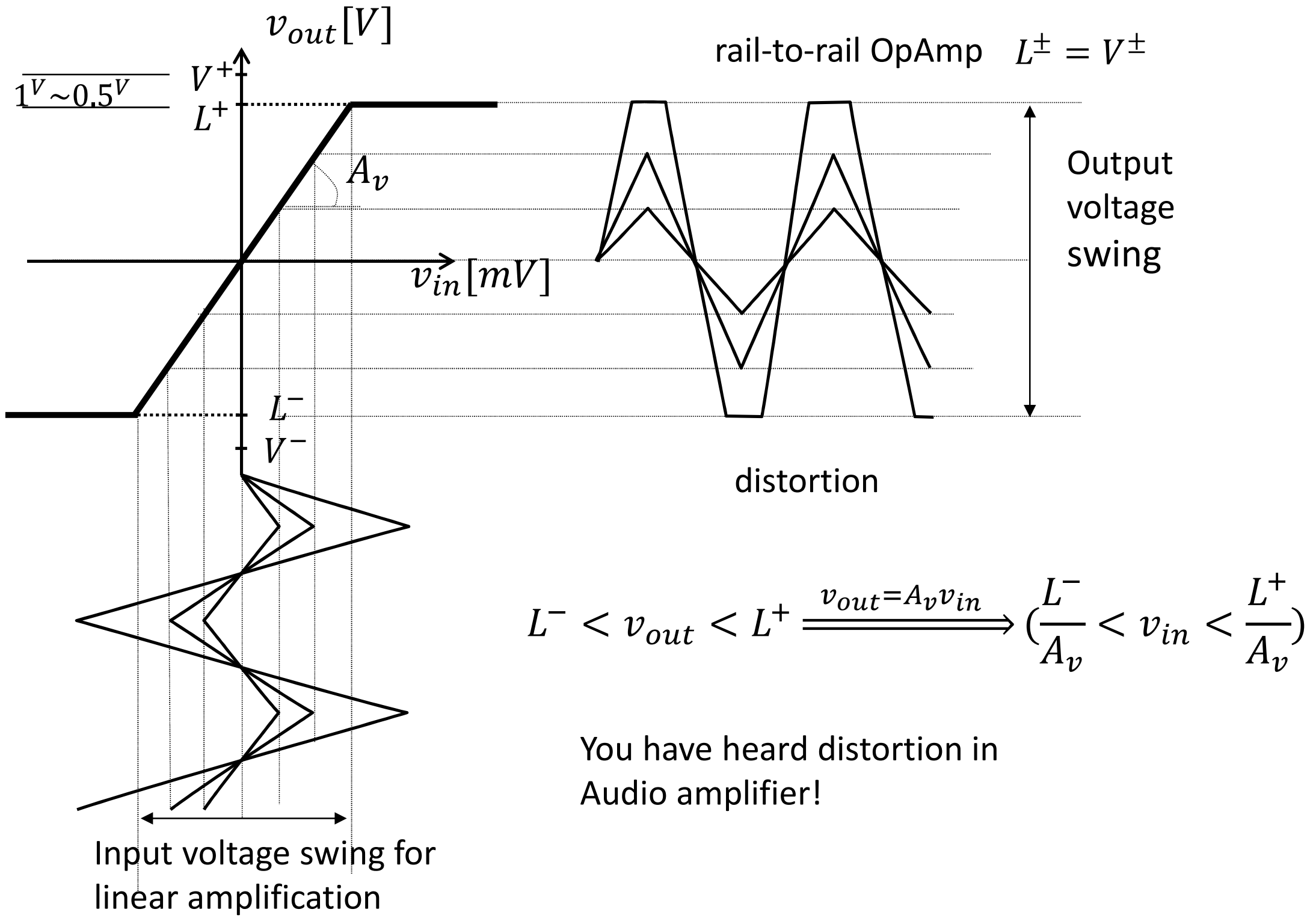


Week2 Electronics1

Amplifiers 2

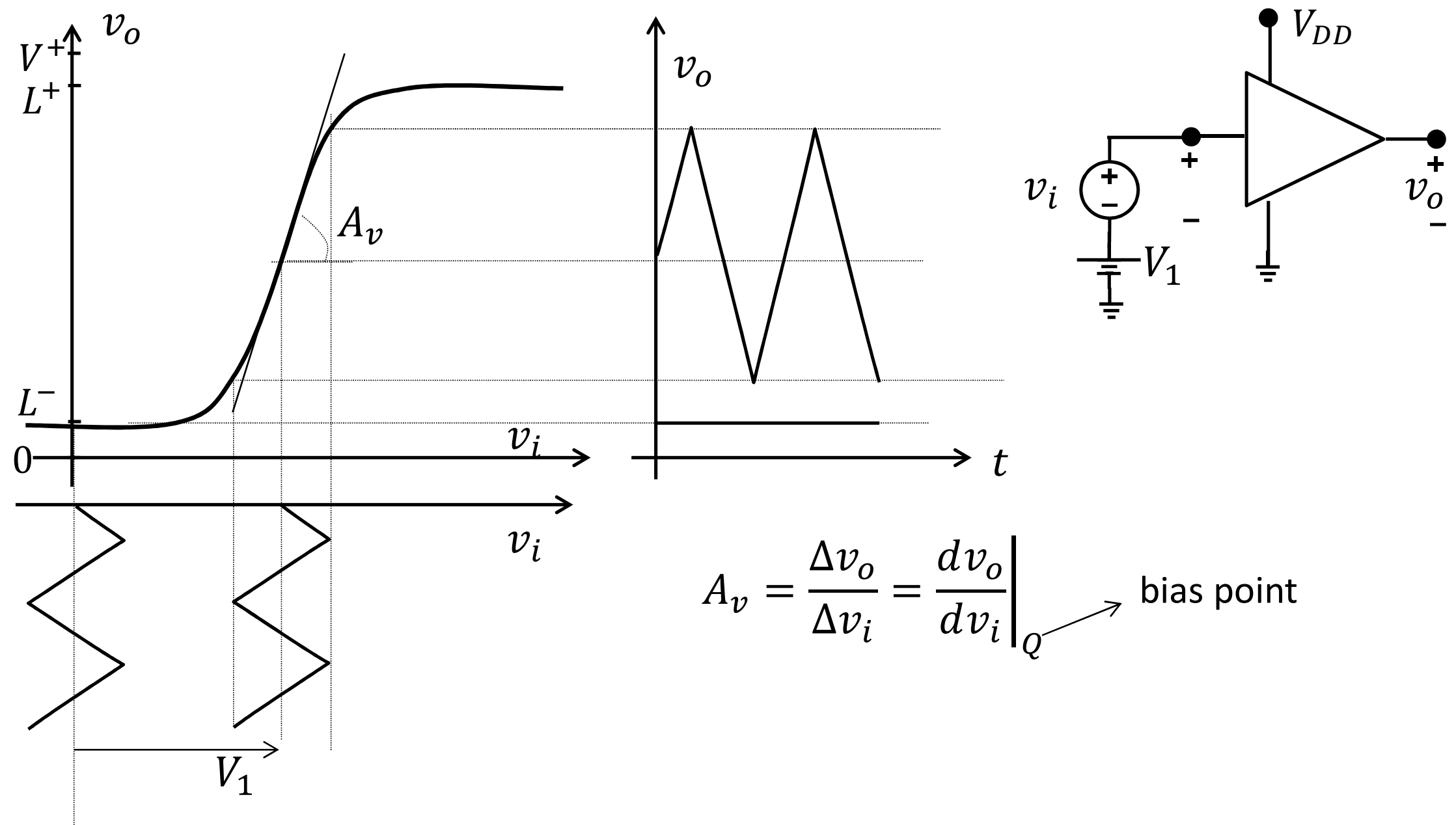


Non-ideal Amplifier



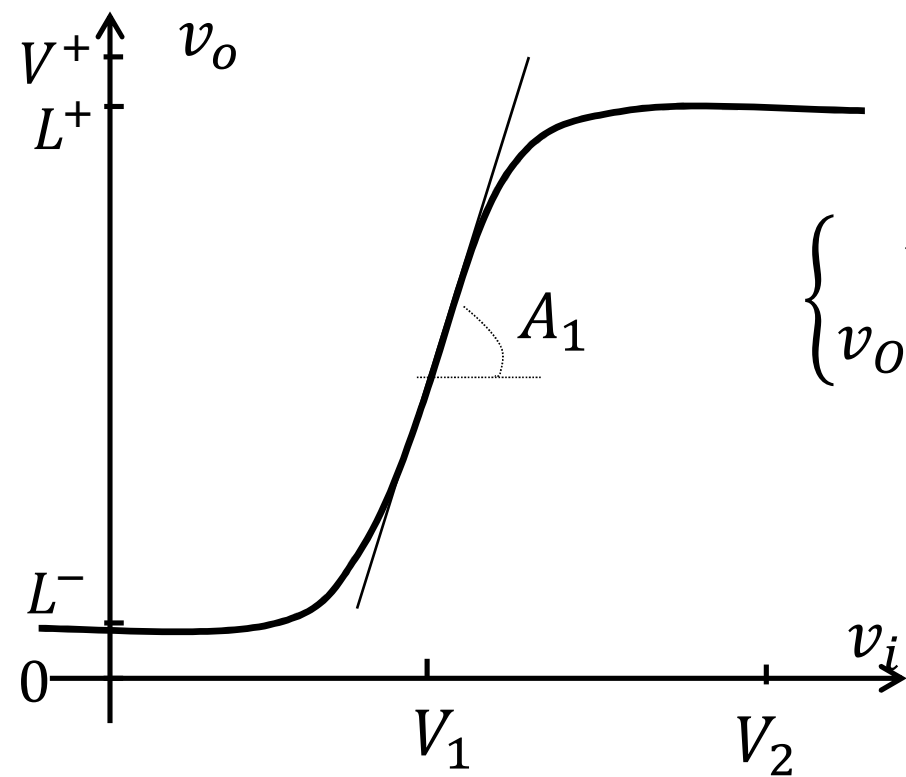
Nonlinear Transfer Function

Biasing

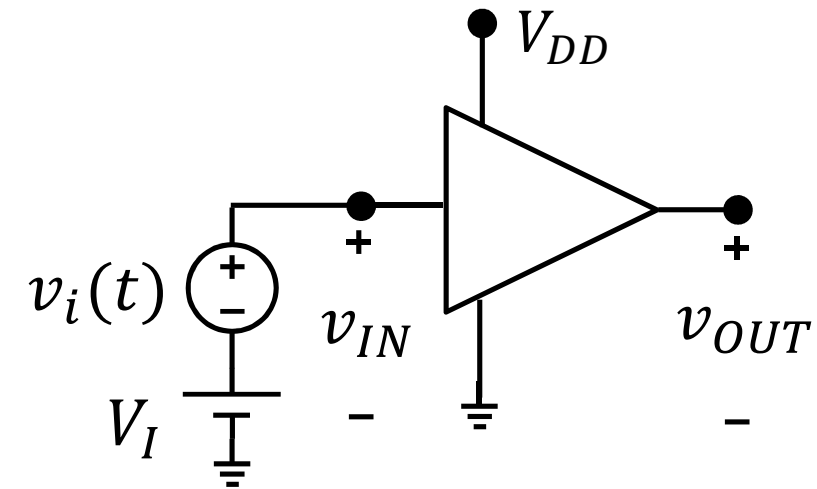


Nonlinear Transfer Function

Biasing



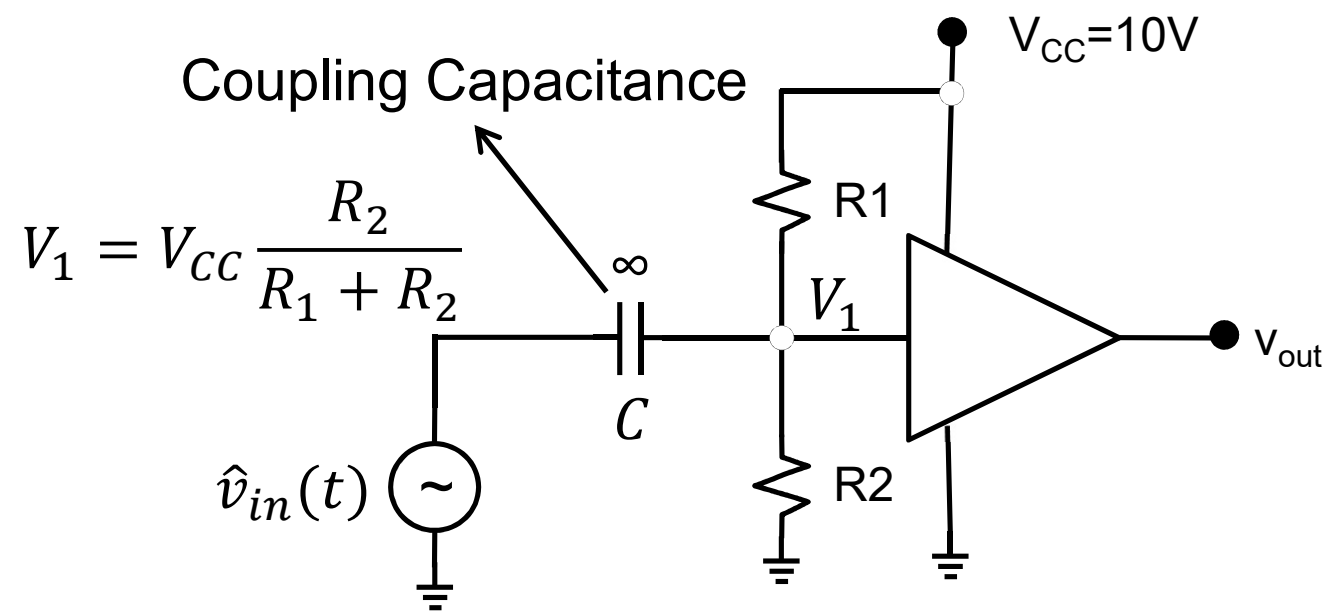
$$\begin{cases} v_{IN} = V_I + v_i(t) \\ v_{OUT} = V_O + A_v v_i(t) \end{cases}$$



$$V_I = 0 \rightarrow A_v = 0$$

$$V_I = V_1 \rightarrow A_v = A_1$$

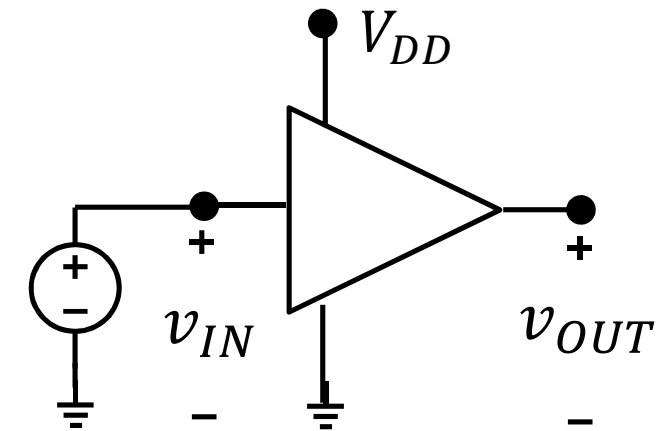
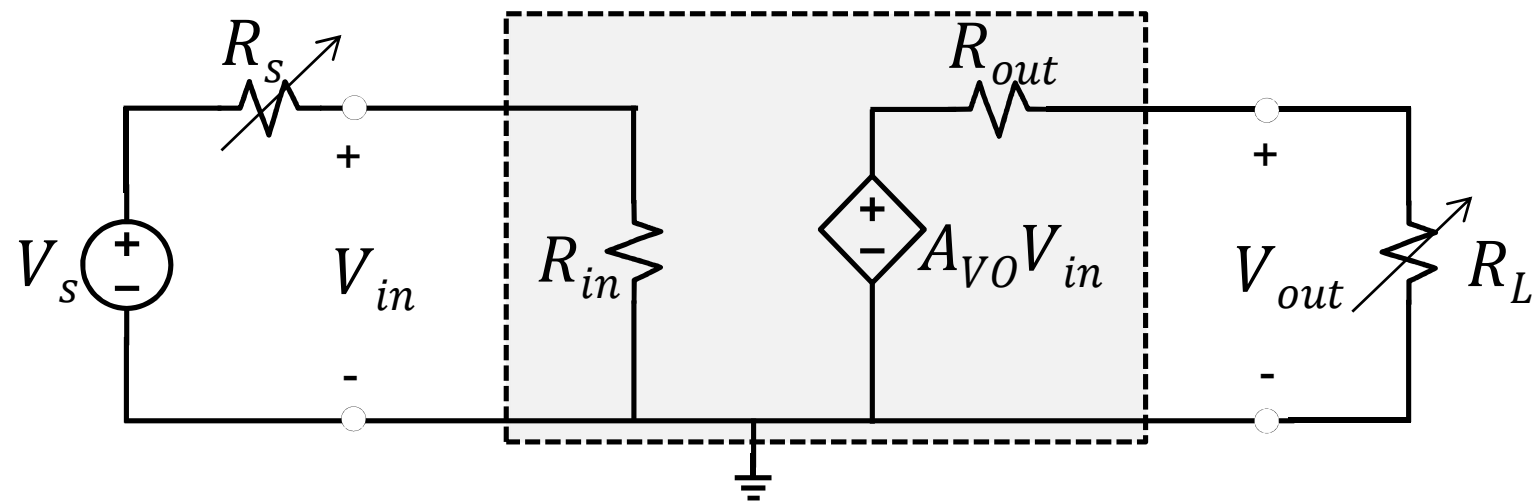
$$V_I = V_2 \rightarrow A_v = 0$$



$$v_{out}(t) = V_{out} + A_v \hat{v}_{in}(t)$$



Voltage Amplifier



Open circuit: $R_L = \infty \rightarrow \frac{V_{out}}{V_{in}} = A_{VO} \left[\frac{V}{V} \right]$

open circuit voltage gain

$$V_{in} = \frac{R_{in}}{R_{in} + R_s} \cdot V_s$$

$$V_{in} = V_s \quad \forall R_s$$

if $R_{in} = \infty$ Ideal case

$$V_{out} = A_{VO} V_{in} \frac{R_L}{R_L + R_{out}}$$

$$V_{out} = A_{VO} V_{in} \quad \forall R_L$$

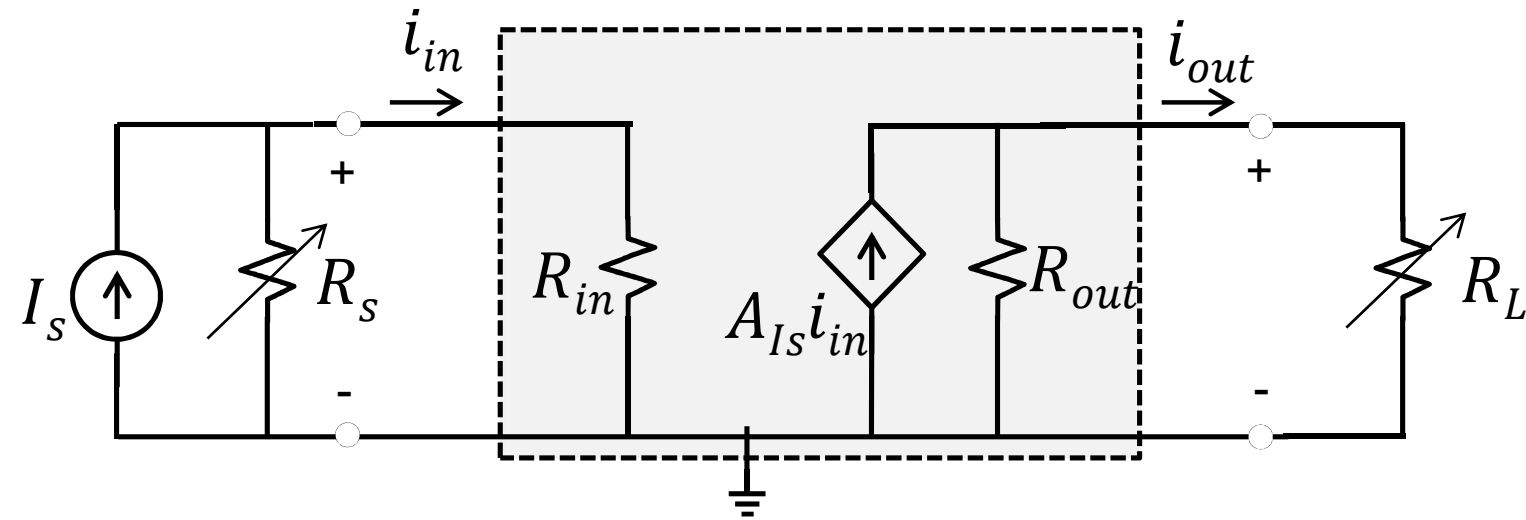
if $R_{out} = 0$ Ideal case

Generally: $\frac{V_{out}}{V_s} = A_{VO} \left(\frac{R_L}{R_L + R_{out}} \right) \left(\frac{R_{in}}{R_{in} + R_s} \right)$

Ideal case: $\frac{V_{out}}{V_s} = A_{VO}$



Current Amplifier



Short circuit: $R_L = 0 \rightarrow \frac{i_{out}}{i_{in}} = A_{Is} \begin{bmatrix} A \\ -A \end{bmatrix}$ short circuit current gain

$i_{in} = \frac{R_s}{R_{in} + R_s} \cdot I_s$ $i_{in} = I_s \quad \forall R_s$ if $R_{in} = 0$ Ideal case

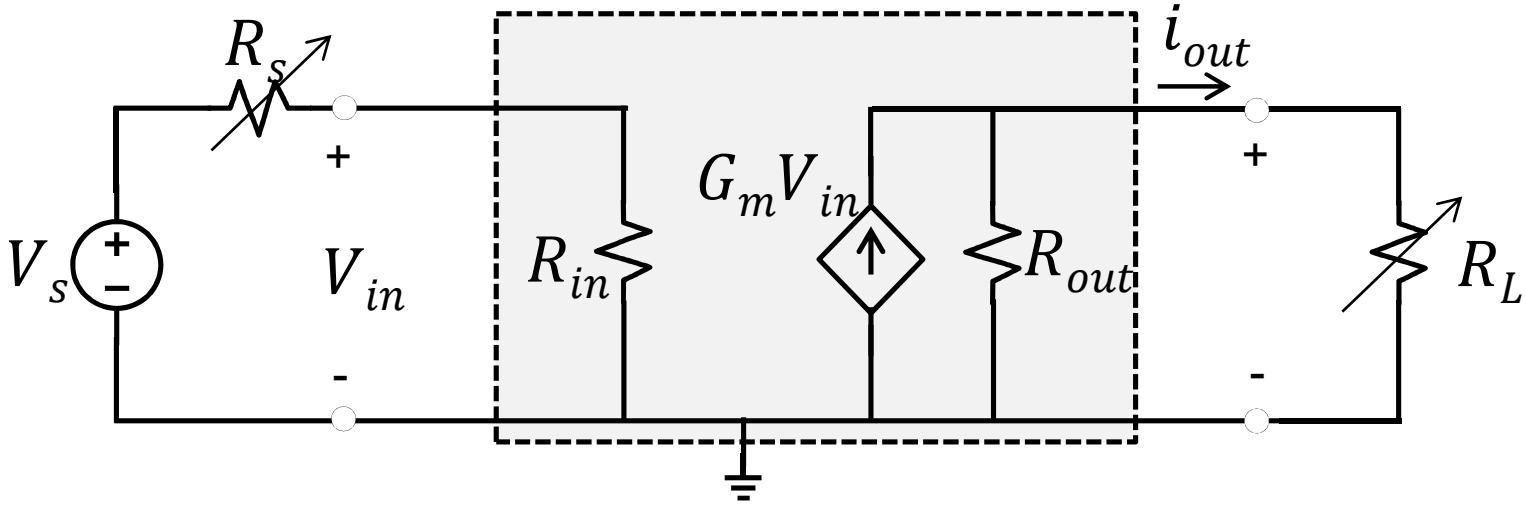
$i_{out} = A_{Is} i_{in} \frac{R_{out}}{R_L + R_{out}}$ $i_{out} = A_{Is} i_{in} \quad \forall R_L$ if $R_{out} = \infty$ Ideal case

Generally: $\frac{i_{out}}{I_s} = A_{Is} \left(\frac{R_{out}}{R_L + R_{out}} \right) \left(\frac{R_s}{R_{in} + R_s} \right)$

Ideal case: $\frac{i_{out}}{I_s} = A_{Is}$



TransConductance Amplifier



Short circuit: $R_L = 0 \rightarrow \frac{i_{out}}{V_{in}} = G_m \quad \left[\frac{A}{V} = \Omega^{-1} \right]$ short circuit Trans-conductance

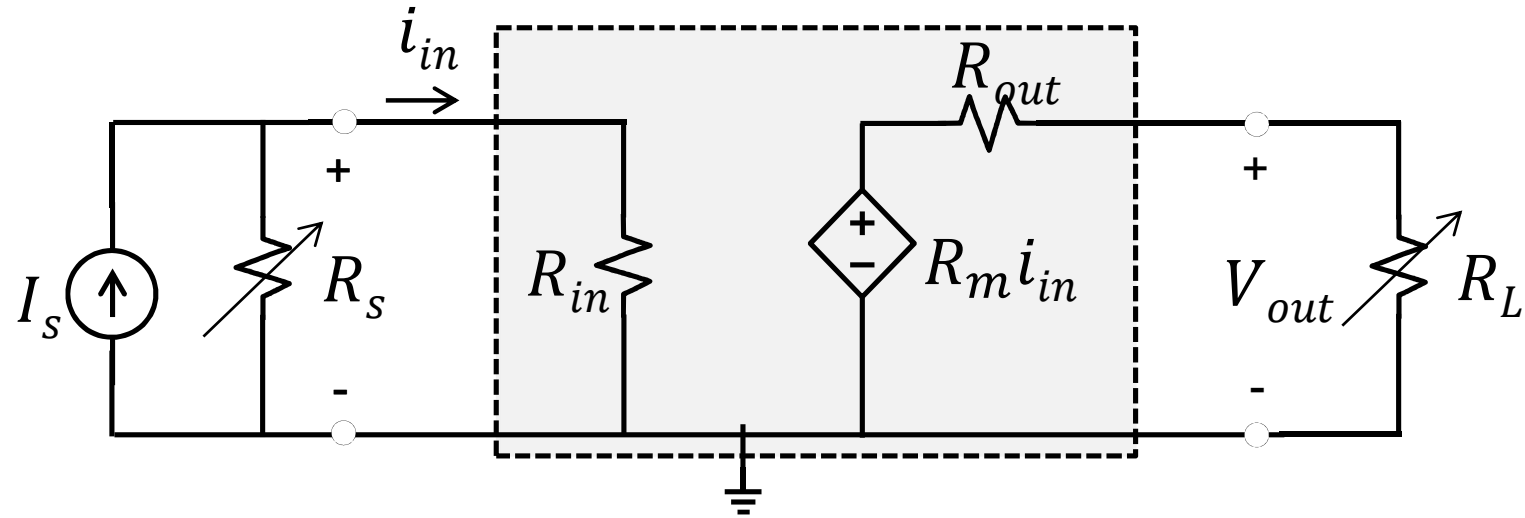
$V_{in} = \frac{R_{in}}{R_{in} + R_s} \cdot V_s$ $V_{in} = V_s \quad \forall R_s$ if $R_{in} = \infty$ Ideal case

$i_{out} = G_m V_{in} \frac{R_{out}}{R_L + R_{out}}$ $i_{out} = G_m V_{in} \quad \forall R_L$ if $R_{out} = \infty$ Ideal case

Generally: $\frac{i_{out}}{V_s} = G_m \left(\frac{R_{out}}{R_L + R_{out}} \right) \left(\frac{R_{in}}{R_{in} + R_s} \right)$

Ideal case: $\frac{i_{out}}{V_s} = G_m$

TransResistance Amplifier



Short circuit: $R_L = \infty \rightarrow \frac{V_{out}}{i_{in}} = R_m \quad \left[\frac{V}{A} = \Omega \right]$ open circuit Trans-resistance

$i_{in} = \frac{R_s}{R_{in} + R_s} \cdot I_s$ $i_{in} = I_s \quad \forall R_s$ if $R_{in} = 0$ Ideal case

$V_{out} = R_m i_{in} \frac{R_L}{R_L + R_{out}}$ $V_{out} = R_m i_{in} \quad \forall R_L$ if $R_{out} = 0$ Ideal case

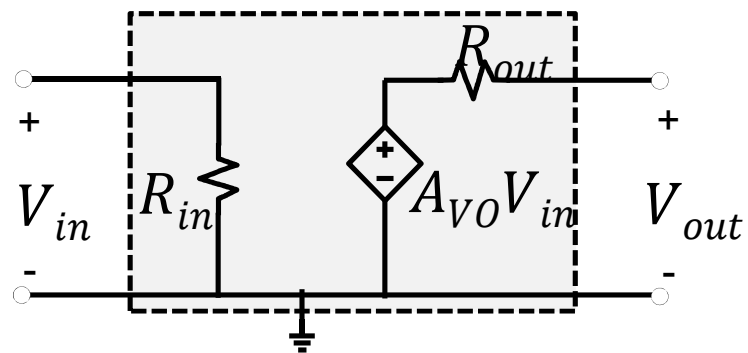
Generally: $\frac{V_{out}}{I_s} = R_m \left(\frac{R_L}{R_L + R_{out}} \right) \left(\frac{R_s}{R_{in} + R_s} \right)$

Ideal case: $\frac{V_{out}}{I_s} = R_m$



Amplifiers

Voltage Amplifier



$$A_{VO} = \left. \frac{V_{out}}{V_{in}} \right|_{i_{out}=0}$$

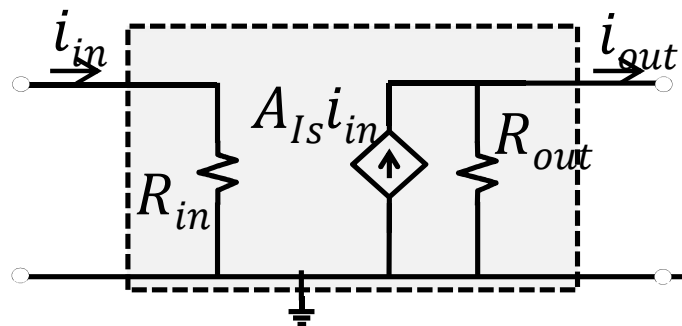
open circuit voltage gain

Ideal:

$$R_{in} = \infty$$

$$R_{out} = 0$$

Current Amplifier



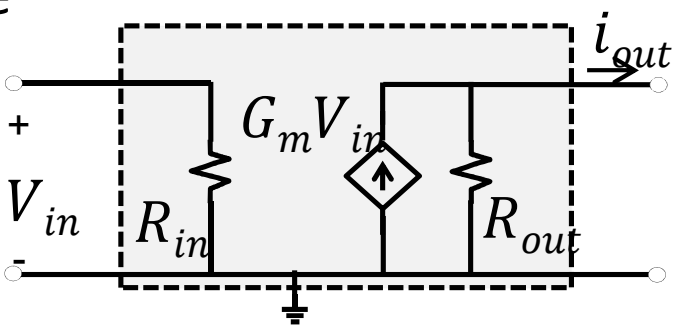
$$A_{IS} = \left. \frac{i_{out}}{i_{in}} \right|_{V_{out}=0}$$

short circuit current gain

$$R_{in} = 0$$

$$R_{out} = \infty$$

Trans-Conductance Amplifier



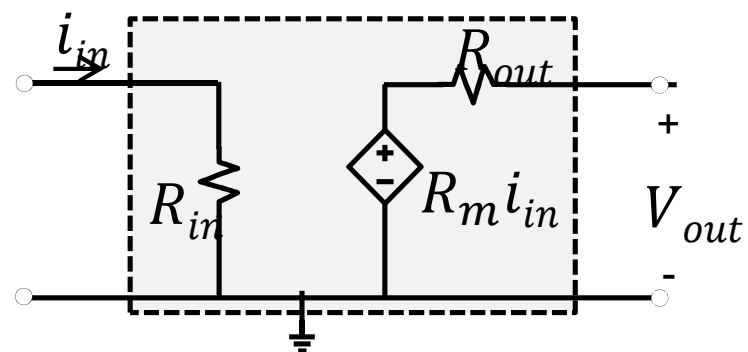
$$G_m = \left. \frac{i_{out}}{V_{in}} \right|_{V_{out}=0}$$

short circuit Trans-conductance

$$R_{in} = \infty$$

$$R_{out} = \infty$$

Trans-Resistance Amplifier



$$R_m = \left. \frac{V_{out}}{i_{in}} \right|_{i_{out}=0}$$

open circuit Trans-resistance

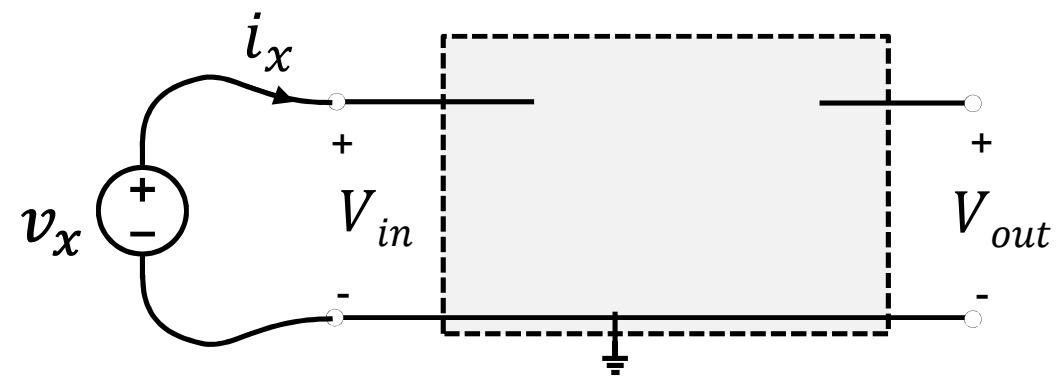
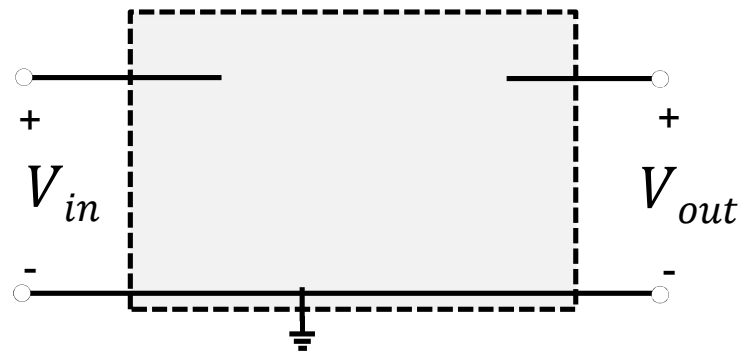
$$R_{in} = 0$$

$$R_{out} = 0$$



Amplifiers

Voltage Amplifier

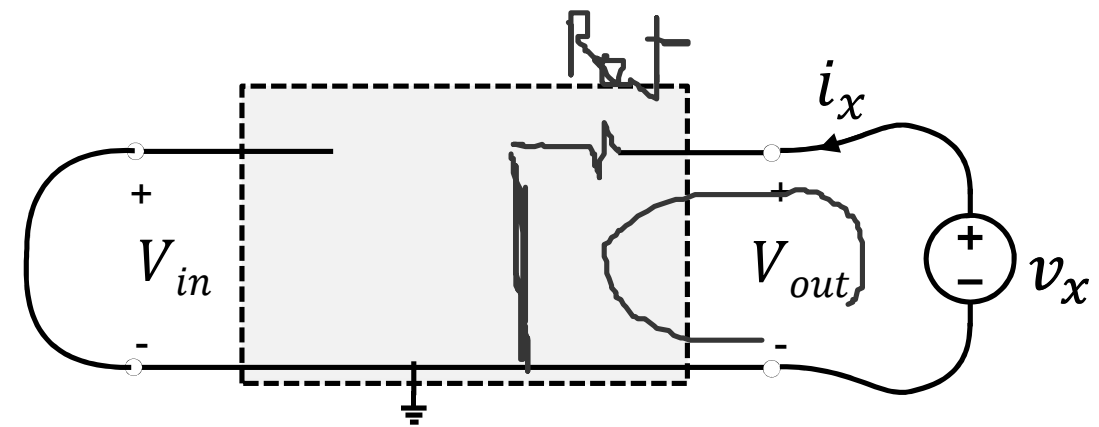
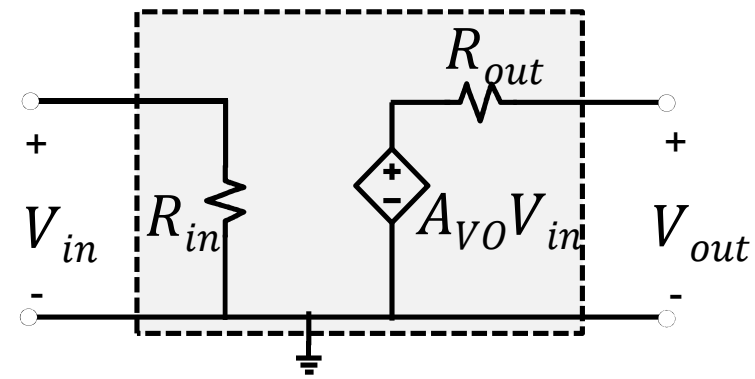


$$R_{in} = \frac{v_x}{i_x}$$

$$V_{in} = v_x$$

$$A_{VO} = \left. \frac{V_{out}}{v_x} \right|_{I_{out}=0}$$

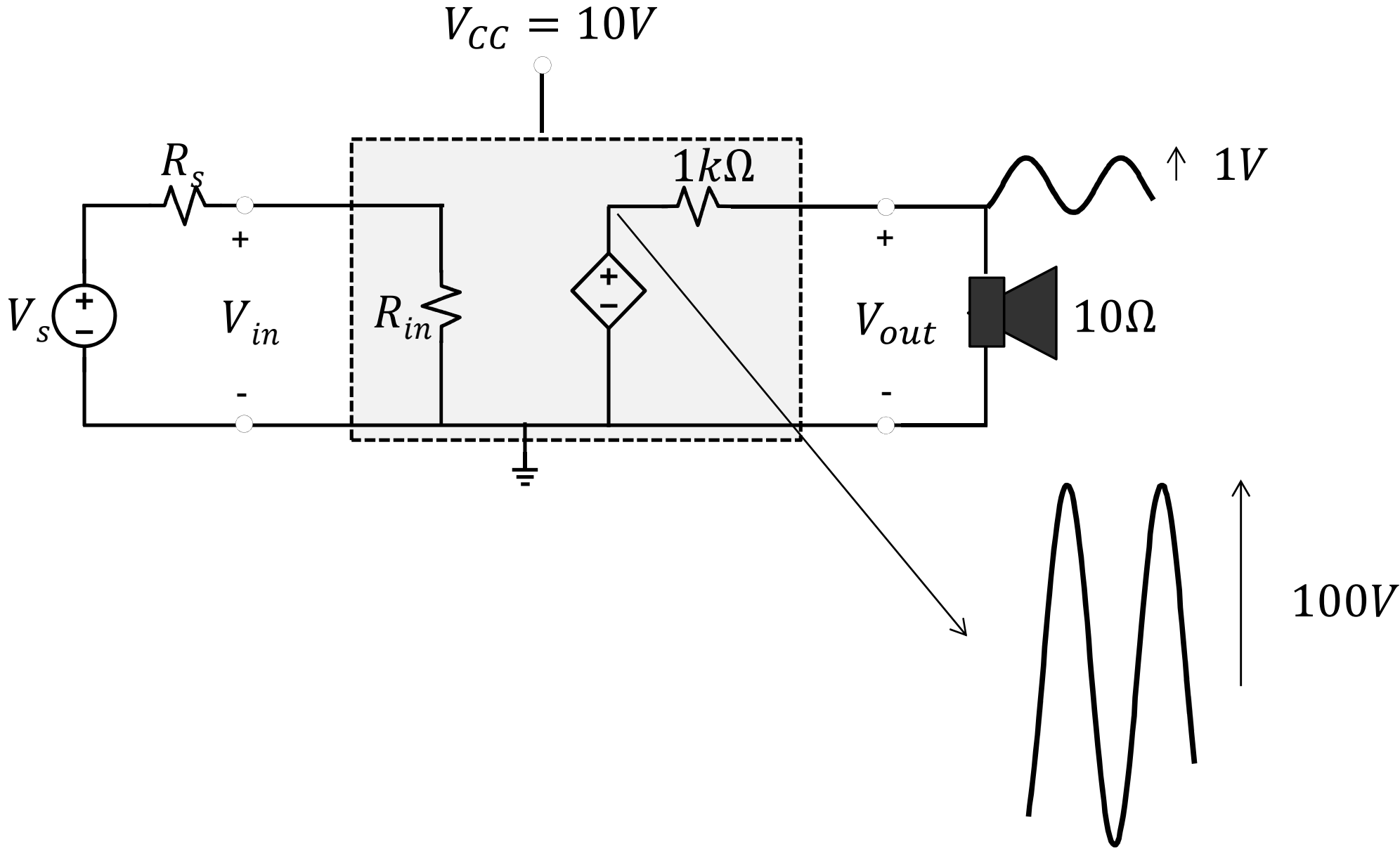
Open Output



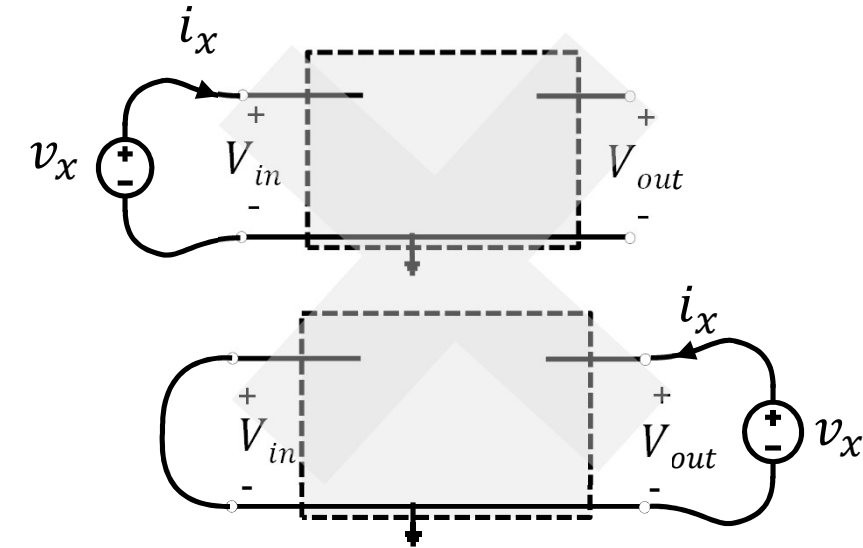
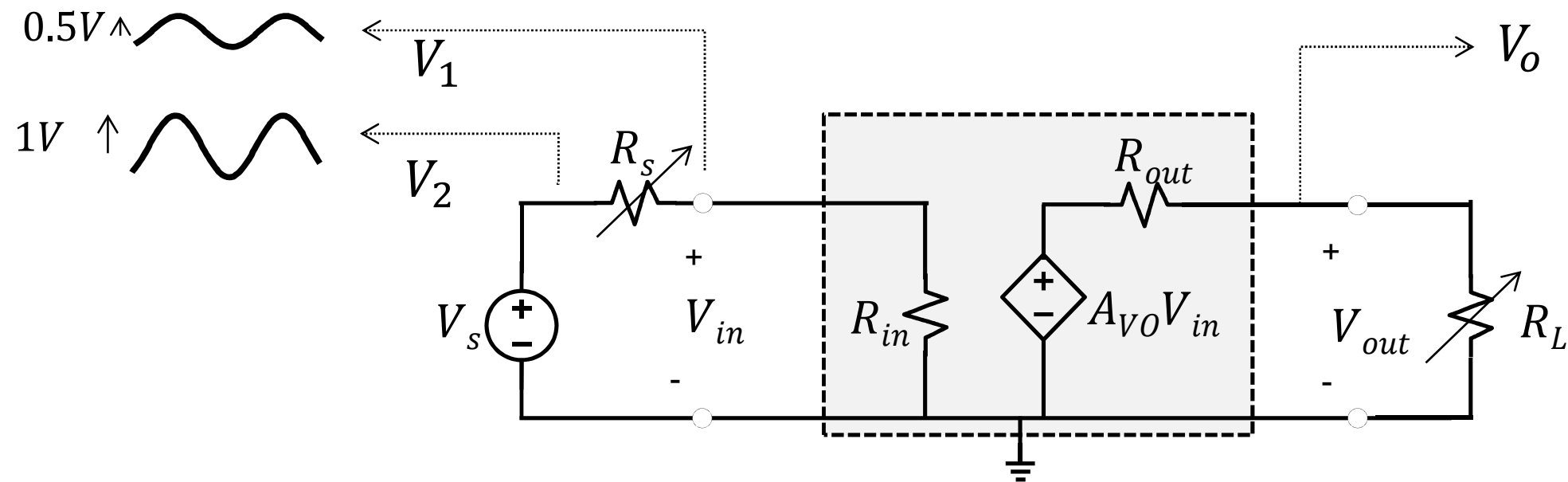
$$R_{out} = \left. \frac{v_x}{i_x} \right|_{V_{in}=0}$$



Practical Consideration



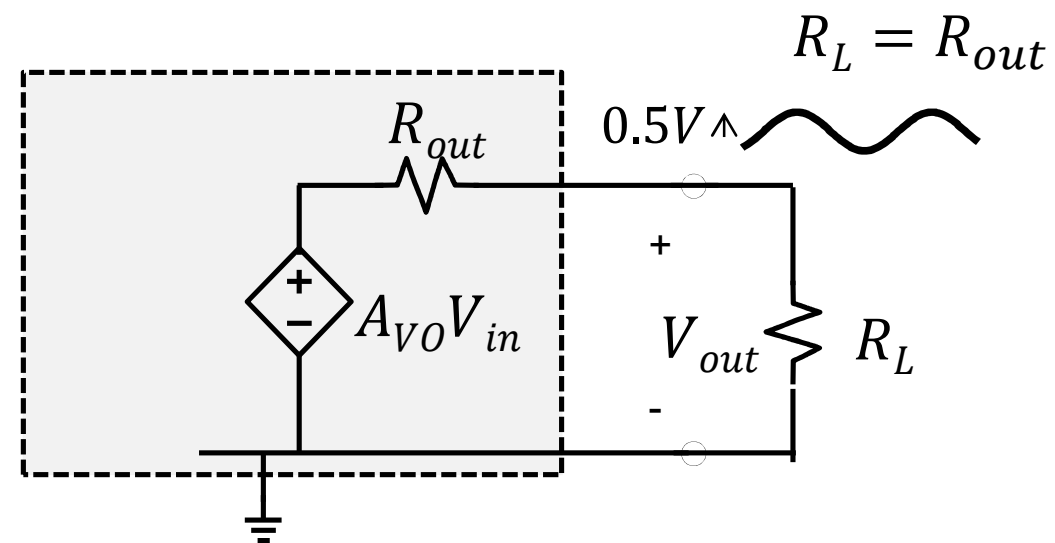
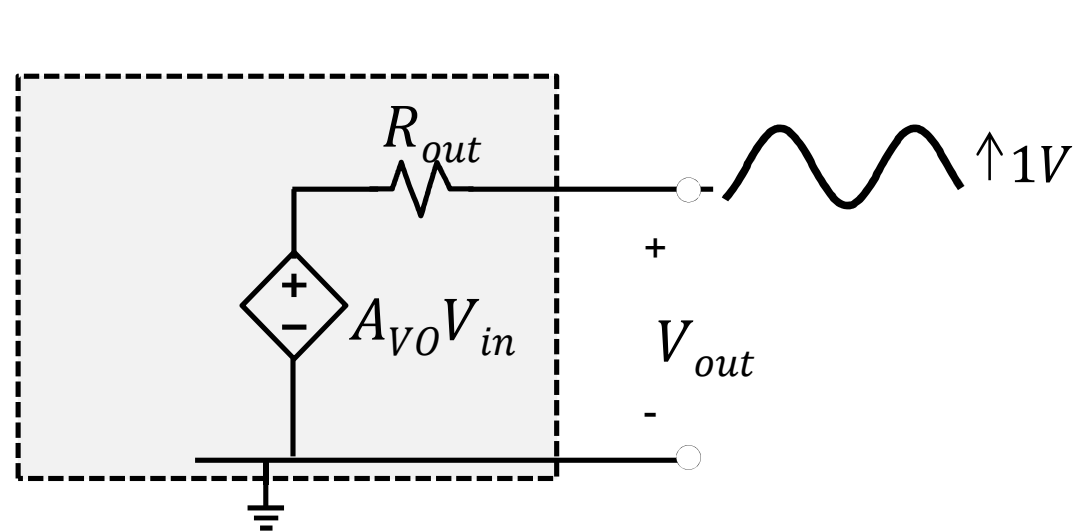
Practical Consideration: Input / Output Resistance



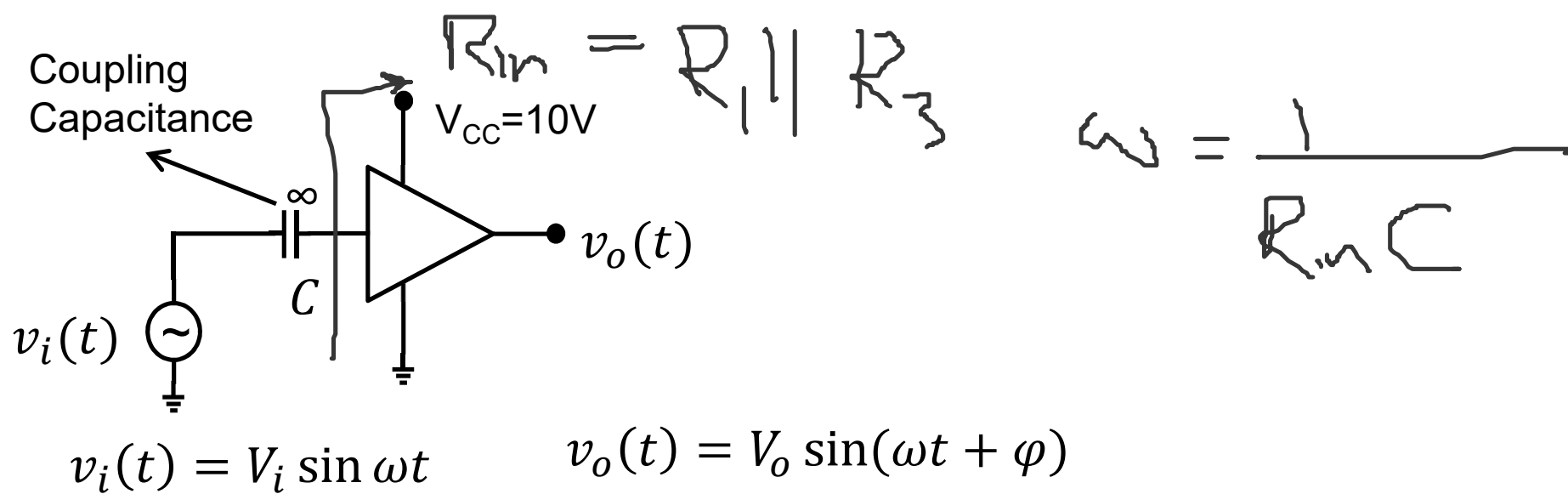
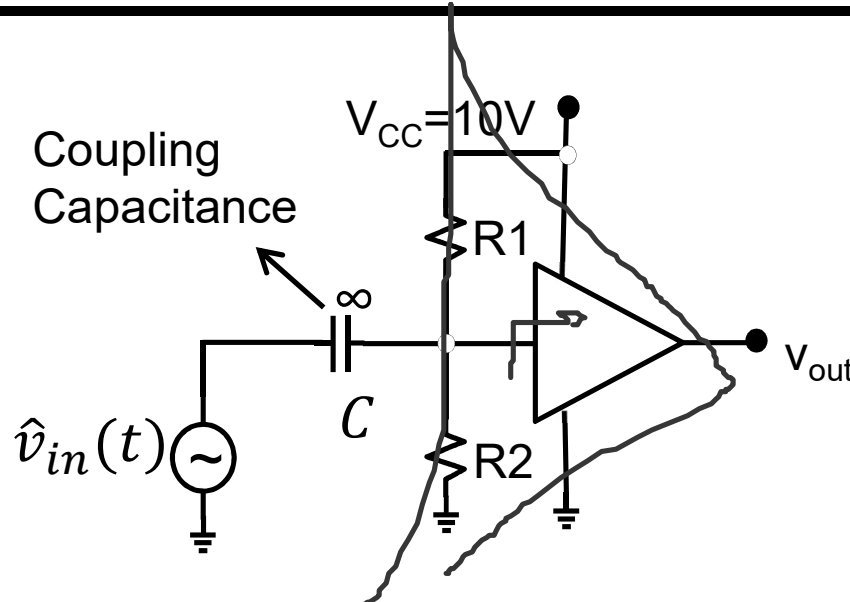
$$R_s = 1k\Omega \rightarrow \frac{V_1}{V_2} = \frac{R_{in}}{R_{in} + 1k} \quad R_{in} = 1k \times \frac{1}{\frac{V_2}{V_1} - 1}$$

Point: You need to make sure circuit is in its linear operation regime (desired bias point)

How about R_{out} ?



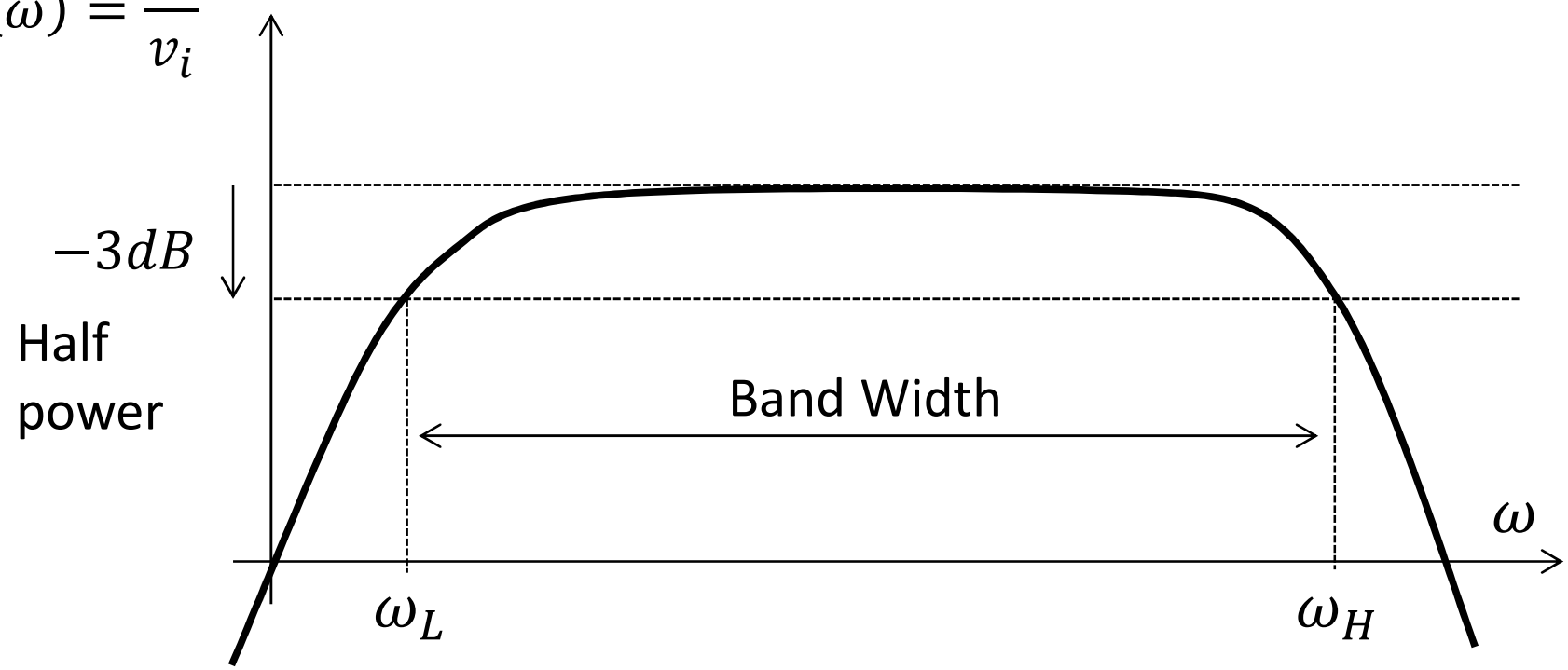
Amplifier Frequency Response



Transfer Function: $T(\omega) = \frac{v_o}{v_i}$

$|T(\omega)| = \frac{V_o}{V_i}$
Amplitude in dB

$\angle T(\omega) = \varphi$
Phase



end
